



Fig. 1.

We see then that in the special case when a square-root step response is desired, the $Z_n(s)$ give a simple and easily realized solution to the time domain synthesis problem. Of more general interest is the fact that $s^{-1/2}$, an irrational function with a branch point at the origin, can be approximated by rational functions whose poles and zeros alternate on the negative-real axis (a branch cut). Whether or not this is a particular example of a more general situation is an open question.

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An RC Impedance Approximant to $s^{-1/2}$ *

Usually, time domain synthesis procedures do not lead easily to realizations. Quite by accident the author came across a method for approximating a \sqrt{t} step response which leads directly to simple RC realizations with known component values. It is hoped

that such an example will be of interest to readers of these TRANSACTIONS.

The method is based on the observation that

$$(1 - s^{-1/2})^n \rightarrow 0 \text{ as } n \rightarrow \infty \quad (1)$$

if $|1 - s^{-1/2}| < 1$. This region of convergence includes the half-plane $\text{Re}\{s\} > 1/4$. When (1) is expanded as a binomial and the terms in $s^{-1/2}$ collected, the following rational approximant is obtained:

$$Z_n(s) = \frac{\sum_{k=0,2,\dots}^{n-1} \binom{n}{k} s^{-k/2}}{\sum_{k=1,3,\dots}^{n-1} \binom{n}{k} s^{-(k-1)/2}} = \frac{\sum_{k=1,3,\dots}^n \binom{n}{k} s^{(k-1)/2}}{\sum_{k=0,2,\dots}^{n-1} \binom{n}{k} s^{k/2}} \rightarrow s^{-1/2}$$

where n is odd. We shall take n to be odd throughout the remainder of this communication. $Z_n(s)$ can also be written

$$Z_n(s) = \left[\frac{(1 + s^{1/2})^n - (1 - s^{1/2})^n}{(1 + s^{1/2})^n + (1 - s^{1/2})^n} \right] s^{-1/2} \quad (2)$$

and the zeros and poles of $Z_n(s)$ are found from this expression to lie on the negative-real axis at the following points:

$$\text{zeros: } \sigma = -\tan^2 k\pi/n$$

$$\text{poles: } \sigma = -1/\tan^2 k\pi/n, \quad k = 1, 2, \dots, (n-1)/2.$$

Therefore, since $Z_n(\infty) = 1/n$,

$$Z_n(s) = \frac{1}{n} \prod_{k=1}^{(n-1)/2} \frac{s + \tan^2 k\pi/n}{s + 1/\tan^2 k\pi/n}$$

By using (2) again to evaluate the residues, $Z_n(s)$ can be expanded in a partial fraction expansion as follows:

$$Z_n(s) = \frac{1}{n} + \frac{2}{n} \sum_{k=1}^{(n-1)/2} \frac{1 + 1/\tan^2 k\pi/n}{s + 1/\tan^2 k\pi/n} \quad (3)$$

Since the poles all lie on the negative-real axis and have positive residues, $Z_n(s)$ is an RC driving-point impedance. $Z_n(s)$ can be realized directly from (3) with $(n+1)/2$ resistors and $(n-1)/2$ capacitors, a total of n elements.

The step response of $Z_n(s)$ can be found by expanding $Z_n(s)/s$ in a partial fraction expansion:

$$Z_n(s)/s = \frac{n}{s} - \frac{2}{n} \sum_{k=1}^{(n-1)/2} \frac{1 + \tan^2 k\pi/n}{s + 1/\tan^2 k\pi/n} \rightarrow s^{-3/2}$$

or, in the time domain,

$$\text{step response} = n - \frac{2}{n} \sum_{k=1}^{(n-1)/2} (1 + \tan^2 k\pi/n) e^{-t/\tan^2 k\pi/n} \quad (4)$$

The step response starts at $1/n$ when $t = 0^+$ and becomes asymptotic to n for large t . It thus can approximate

$$\mathcal{L}^{-1}[s^{-3/2}] = 2(t/\pi)^{1/2} \quad (5)$$

well only for $t < \pi n^2/4$.

Fig. 1 shows the over-all behavior of the first few approximations to a $2\sqrt{t/\pi}$ step response. The accuracy in the range of approximation increases quite rapidly. For example, if the error $e(t)$ is defined as the difference between (5) and (4), we have

$$n = 3: \quad |e(t)| \leq .07 \quad \text{for } 0.1 \leq t \leq 5.0$$

$$n = 5: \quad |e(t)| \leq .03 \quad \text{for } 0.1 \leq t \leq 10.0$$

$$n = 7: \quad |e(t)| \leq .011 \quad \text{for } 0.2 \leq t \leq 14.5$$

$$n = 9: \quad |e(t)| \leq .004 \quad \text{for } 0.2 \leq t \leq 19.5.$$

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